

# Control Systems

9

## "State Equations" (part II)

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# State Equations

Recollecting (from part I),  
the *state equations* have the form (systems of  $n^{th}$  order)

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A} \mathbf{x} + \mathbf{B} \mathbf{u} \\ \mathbf{y} = \mathbf{C} \mathbf{x} + \mathbf{D} \mathbf{u} \end{cases}$$

where:

$\mathbf{A}$  is a  $n \times n$  matrix

$\mathbf{B}$  is a  $n \times p$  matrix

$\mathbf{C}$  is a  $q \times n$  matrix

$\mathbf{D}$  is a  $q \times p$  matrix

with:

$p$  = number of **inputs**

$q$  = number of **outputs**

$\dot{\mathbf{x}}$  = **derivative** of vector  $\mathbf{x}$

# State Equations

$\dot{\mathbf{x}}$  = the derivative of vector  $\mathbf{x}$

$$\mathbf{x}(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_{n-1}(t) \\ x_n(t) \end{bmatrix}$$

$$\dot{\mathbf{x}}(t) = \begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \\ \vdots \\ \dot{x}_{n-1}(t) \\ \dot{x}_n(t) \end{bmatrix}$$

## State Equations

For the case of *systems with only one input*  $u(t)$ , i.e.,  $p = 1$ , we have that:

$$B = \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{bmatrix}$$

that is, in this case

$B$  is a *column vector*.

For the case of *systems with only one output*  $y(t)$ , i.e.,  $q = 1$ , we have that:

$$C = [c_1 \quad c_2 \quad \dots \quad c_n]$$

$C$  is a *row vector*.

For the case of systems with only one input  $u(t)$  and one output  $y(t)$ ,

$$D = [d_1]$$

$D$  is a *constant*  $d_1$  (that is,  $D$  is a  $1 \times 1$  matrix).

## State Equations

A matrix  $A$  in the “companion form” has the following aspect:

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots & 0 \\ 0 & 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & \dots & 1 \\ \left(\frac{-a_n}{a_o}\right) & \left(\frac{-a_{n-1}}{a_o}\right) & \left(\frac{-a_{n-2}}{a_o}\right) & \left(\frac{-a_{n-3}}{a_o}\right) & \dots & \left(\frac{-a_1}{a_o}\right) \end{bmatrix}$$

where,  $a_o$ ,  $a_1$ ,  $\dots$ ,  $a_{n-1}$ ,  $a_n$ , are the *coefficients* of the *characteristic equation*  $p(s)$ :

$$p(s) = a_o s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n$$

## State Equations

In the particular case, but very common, of  $a_0 = 1$ , matrix  $A$  in the “companion form” has the following aspect:

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 & \cdots & 0 \\ 0 & 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & \cdots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & -a_{n-3} & \cdots & -a_1 \end{bmatrix}$$

where,  $a_1, \dots, a_{n-1}, a_n$ , are the *coefficients* of the *characteristic equation*  $p(s)$ :

$$p(s) = s^n + a_1 s^{n-1} + a_2 s^{n-2} + \cdots + a_{n-1} s + a_n$$

The *characteristic equation* and the *poles* of the system

A system described in the form of *state equations*

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A} \mathbf{x} + \mathbf{B} \mathbf{u} \\ \mathbf{y} = \mathbf{C} \mathbf{x} + \mathbf{D} \mathbf{u} \end{cases}$$

has its *characteristic polynomial* given by:

$$p(s) = \det \{ [s\mathbf{I} - \mathbf{A}] \}$$

The *poles* of the system are the “*eigenvalues*” of  $A$ , which can be *repeated*, i.e., *double*, *triple*, etc.

It is well known that the *eigenvalues* of  $A$  are the *roots* of the *characteristic polynomial*

$$p(s) = \det [ s \cdot I - A ]$$



## Equivalent representations

Consider a system described by *state equations*

$$\begin{cases} \dot{\mathbf{x}} = \mathbf{A} \mathbf{x} + \mathbf{B} \mathbf{u} \\ \mathbf{y} = \mathbf{C} \mathbf{x} + \mathbf{D} \mathbf{u} \end{cases}$$

which the *state variable* is  $\mathbf{x}(t)$

Now defining a new *state variable*  $\bar{\mathbf{x}}$  as:

$$\bar{\mathbf{x}} = \mathbf{P} \mathbf{x} \quad \mathbf{P} \text{ being } \underline{\text{invertible}}.$$

thus, since:

$$\dot{\bar{\mathbf{x}}} = \mathbf{P} \dot{\mathbf{x}}$$

we have that:

$$\begin{cases} \mathbf{x} = \mathbf{P}^{-1} \bar{\mathbf{x}} \\ \dot{\mathbf{x}} = \mathbf{P}^{-1} \dot{\bar{\mathbf{x}}} \end{cases}$$

# State Equations

that is:

$$\begin{cases} \dot{\bar{x}} = \bar{A} \bar{x} + \bar{B} u \\ y = \bar{C} \bar{x} + \bar{D} u \end{cases}$$

this is another representation of the same system in *state equations*

where:

$$\bar{A} = P A P^{-1}$$

$$\bar{B} = P B$$

$$\bar{C} = C P^{-1}$$

$$\bar{D} = D$$

Note that the *input*  $u$  and the *output*  $y$  do not change.

Only the *internal representation* of the system (as *state variable*).

## Conversion from the State Equation to Transfer Function

In order to convert the representation of a system  
in *state equations*

$$\begin{cases} \dot{x} = Ax + Bu \\ y = Cx + Du \end{cases}$$

to *transfer function*, the expression is given by,

$$\frac{Y(s)}{U(s)} = C \cdot (sI - A)^{-1} \cdot B + D$$

**analog simulation**

## Analog Simulation

No matter what is the nature of a *linear time invariant* (LTI) system (mechanical, electrical, electromechanical, thermic, hydraulic, or a chemical process, etc.) it can be simulated in laboratory by using electronic components.

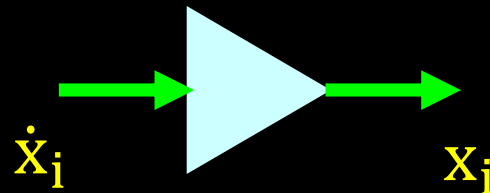
In that way it is possible to simulate the *input* of any *system*, such as for example a step function, and to observe what it would be the *output* of the *system* for that *input*.

That is called “*analog simulation*”.

# State Equations

Components with which we do the *analog simulation*

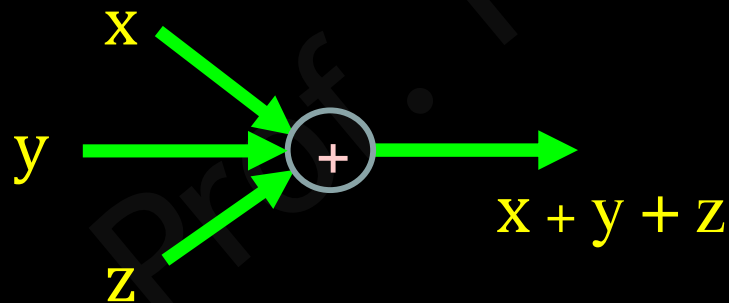
## INTEGRATOR



transforms an **input** signal  $\dot{x}_i$  in  $x_i$  in its **output**, that is, it **integrates**

The *analog simulation* of a  $n$  order system needs  $n$  integrators.

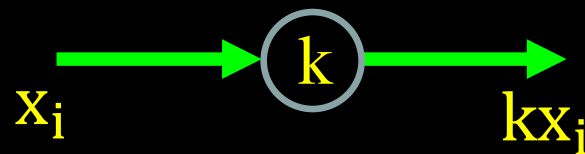
## ADDING ELEMENT



**sums** the **input** signals giving an **output** signal

# State Equations

## MULTIPLIER

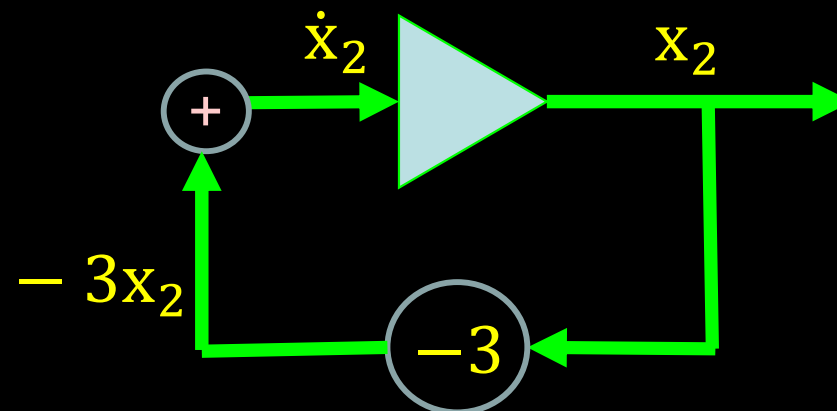


multiplies the *input* signal  $x_i$  by  $k$ , giving back  $kx_i$  in its *output*

## Example 16:

In the figure below we can see how it is done the *analog simulation* of the *differential equation*

$$\dot{x}_2 = -3x_2$$

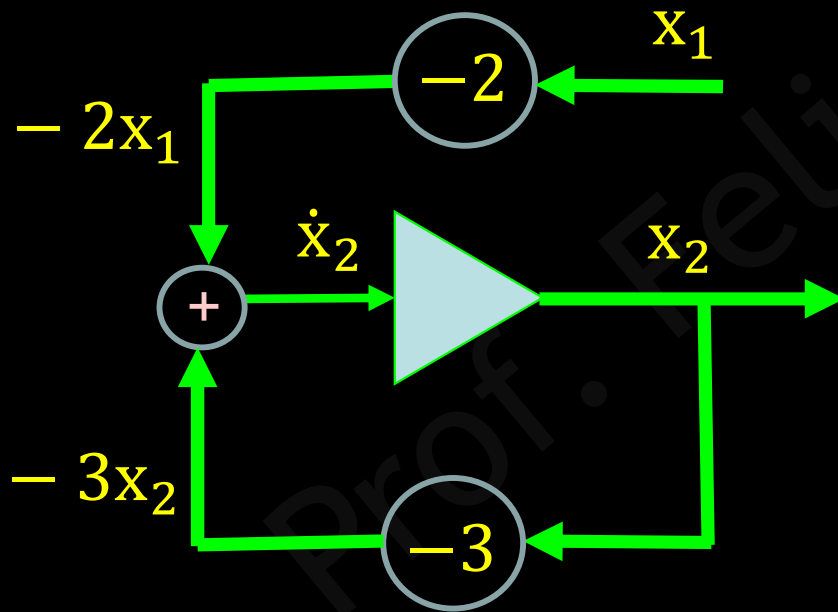


# State Equations

## Example 17:

Now the *analog simulation* of the differential equation

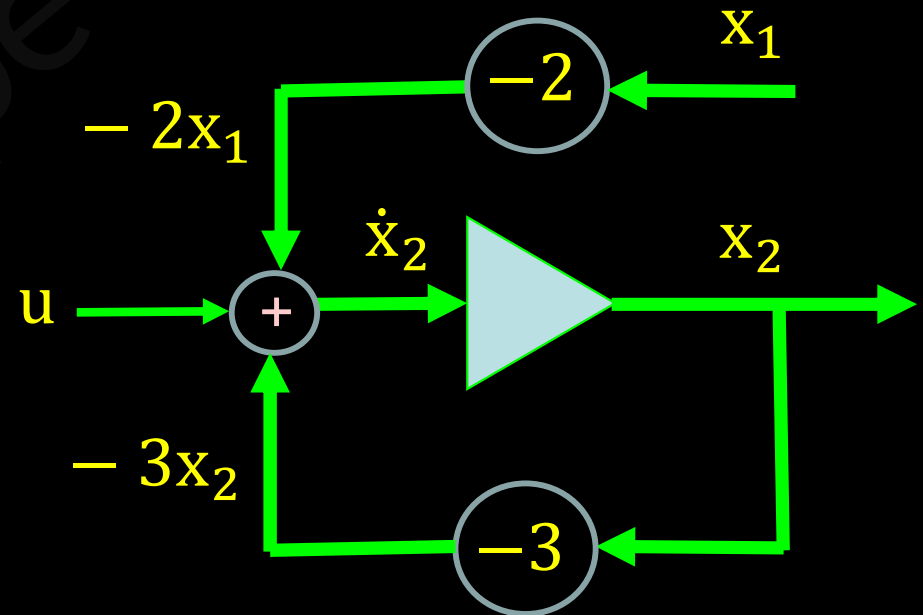
$$\dot{x}_2 = -2x_1 - 3x_2$$



## Example 18:

And now the *analog simulation* of the differential equation

$$\dot{x}_2 = -2x_1 - 3x_2 + u$$



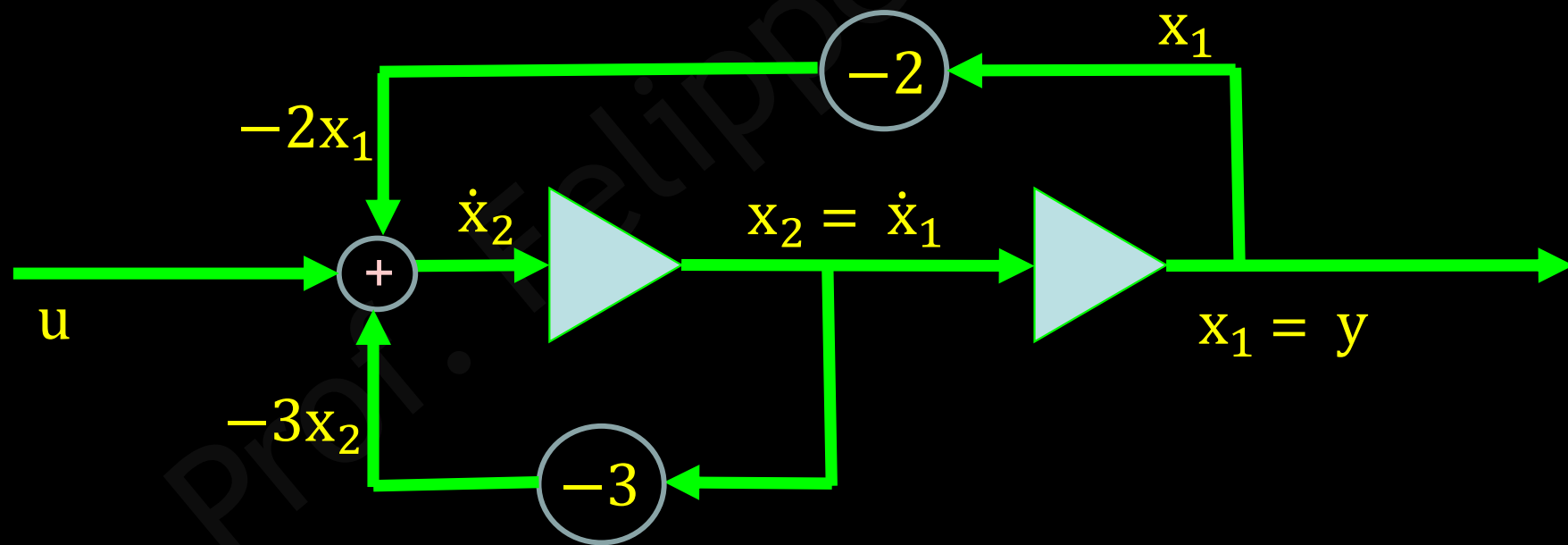


# State Equations

## Example 19:

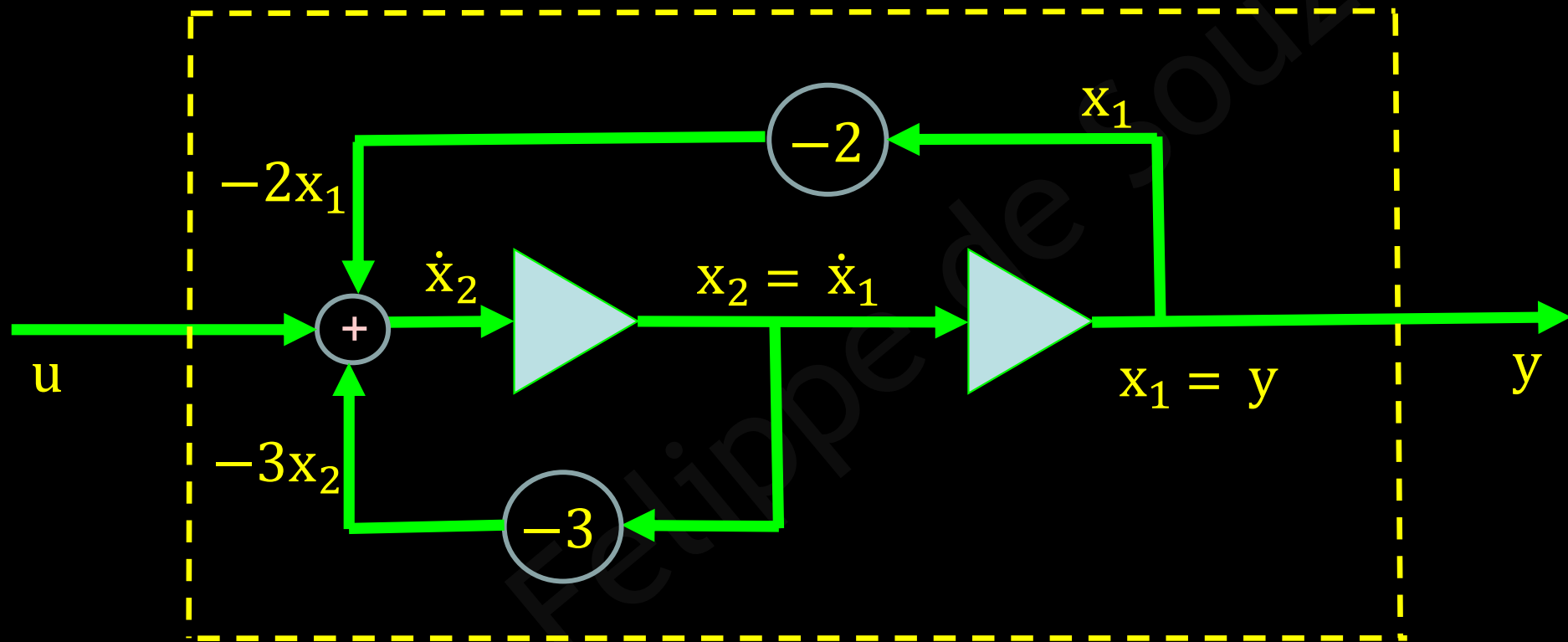
$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = -2x_1 - 3x_2 + u \\ y = x_1 \end{cases}$$

Let us now do the *analog simulation* of this 2<sup>nd</sup> order system (described by its *state equations*)



## Example 19 (continued):

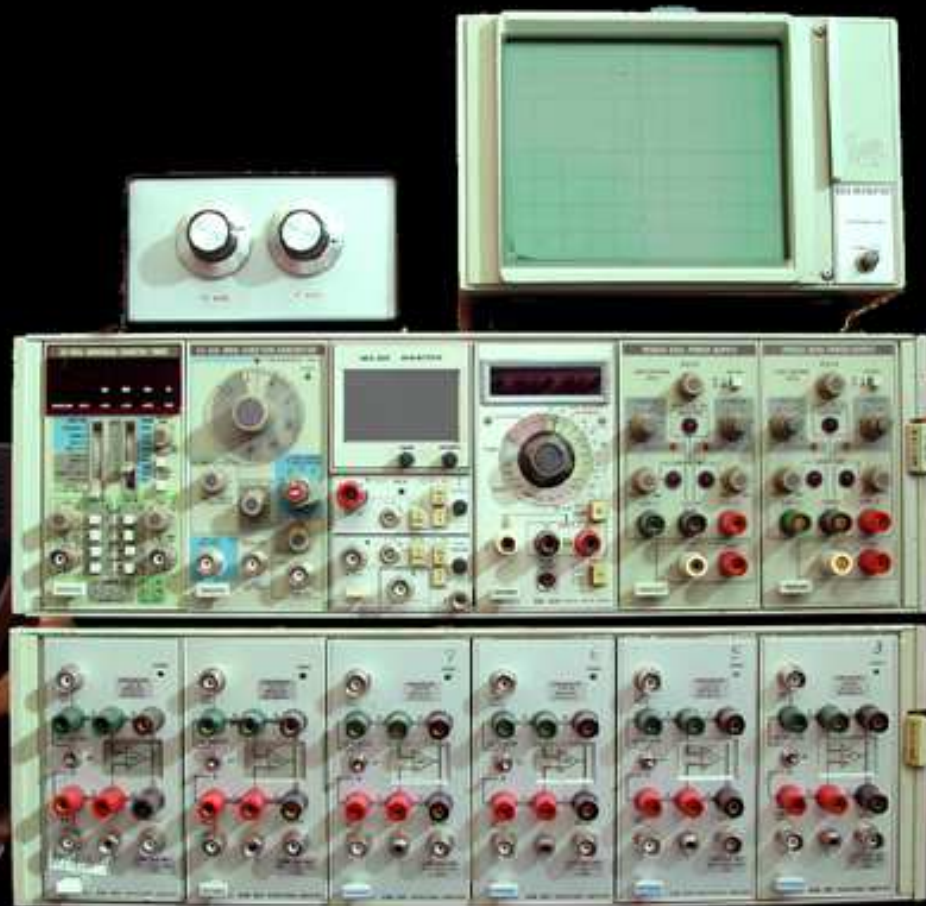
Now, if we draw a box covering the *analog simulation* done



we can observe that in this box only the *input*  $u$  comes in and only the *output*  $y$  comes out.

The variables  $x_1$  and  $x_2$  which stayed inside the box are part of an internal representation of the *system*, though its *state variable*

## Analog Simulation in practice



**conversion of the transfer function  
to state equations**

## Conversion of the Transfer Function to State Equations

We have seen the representation of a system by its *transfer function*

$$\frac{Y(s)}{U(s)} \text{ is unique!}$$

On the other hand the representation of a system in *state equations*

$$\begin{cases} \dot{x} = A x + B u \\ y = C x + D u \end{cases}$$

It is not unique!

## State Equations

There is no single rule to transform systems described by its *ordinary differential equation* (ODE) or by its *transfer function*, in *state equations*

Let us see the same *third order* system A, described by the *differential equation*

$$\ddot{y} + 12\dot{y} + 20y = 80u$$

which its *transfer function* is given by

$$G(s) = \frac{Y(s)}{U(s)} = \frac{80}{s^3 + 12s^2 + 20s}$$

or

$$G(s) = \frac{Y(s)}{U(s)} = \frac{80}{s(s + 2)(s + 10)}$$

## Example 20:

For system A described by the *ordinary differential equation* (ODE) is given by (1)

$$\ddot{y} + 12\dot{y} + 20y = 80u$$

Defining the  
*state variable*

$$\begin{cases} x_1 = y \\ x_2 = \dot{y} \\ x_3 = \ddot{y} \end{cases}$$

we get the  
*state equations*

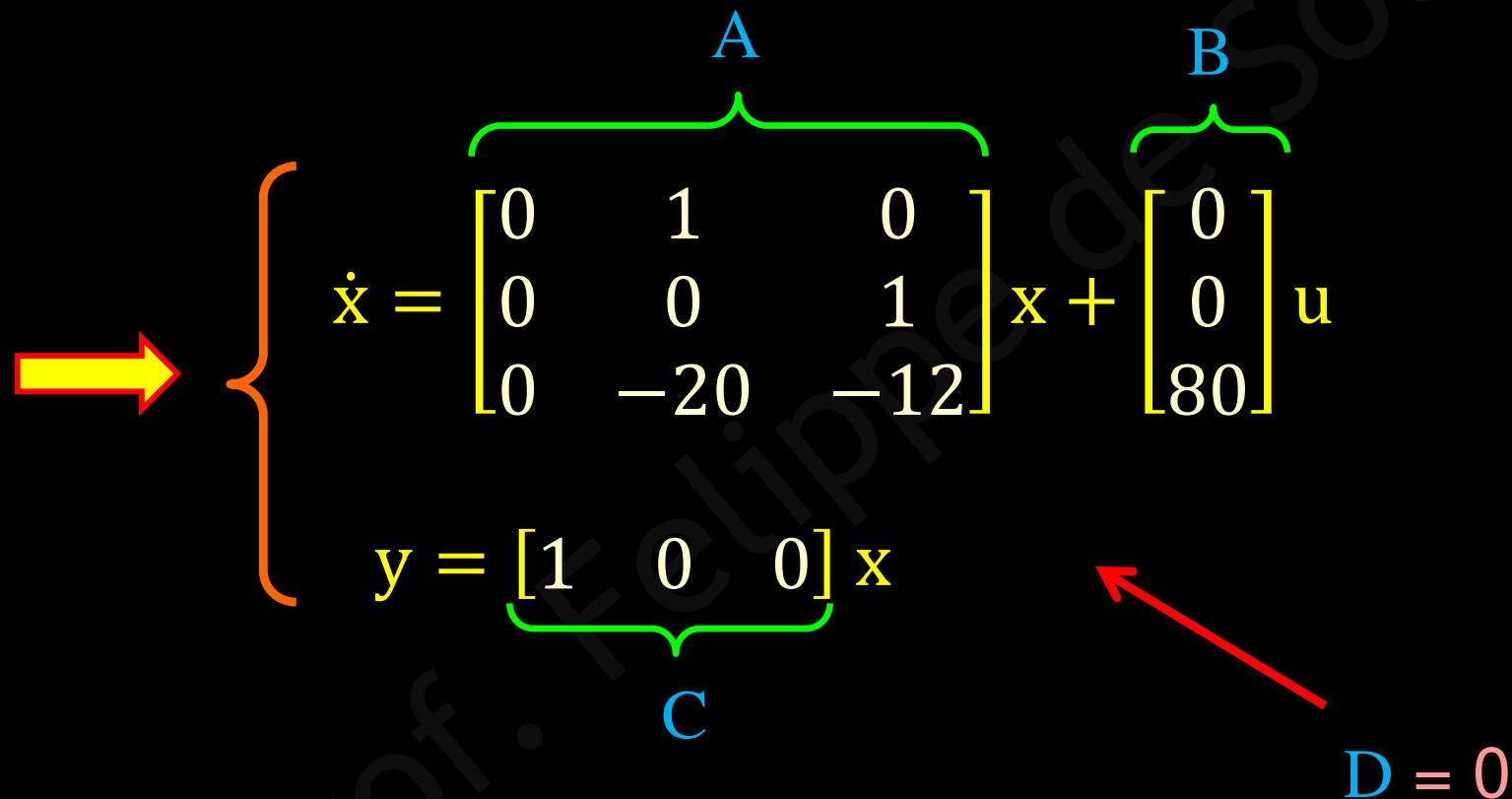


$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = x_3 \\ \dot{x}_3 = -20x_2 - 12x_3 + 80u \\ y = x_1 \end{cases}$$

# State Equations

Example 20 (continued):

and writing in the **matrix form** we get


$$\begin{aligned} \dot{\mathbf{x}} &= \overbrace{\begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & -20 & -12 \end{bmatrix}}^{\mathbf{A}} \mathbf{x} + \overbrace{\begin{bmatrix} 0 \\ 0 \\ 80 \end{bmatrix}}^{\mathbf{B}} u \\ y &= \underbrace{\begin{bmatrix} 1 & 0 & 0 \end{bmatrix}}_{\mathbf{C}} \mathbf{x} \end{aligned}$$

$\mathbf{D} = 0$

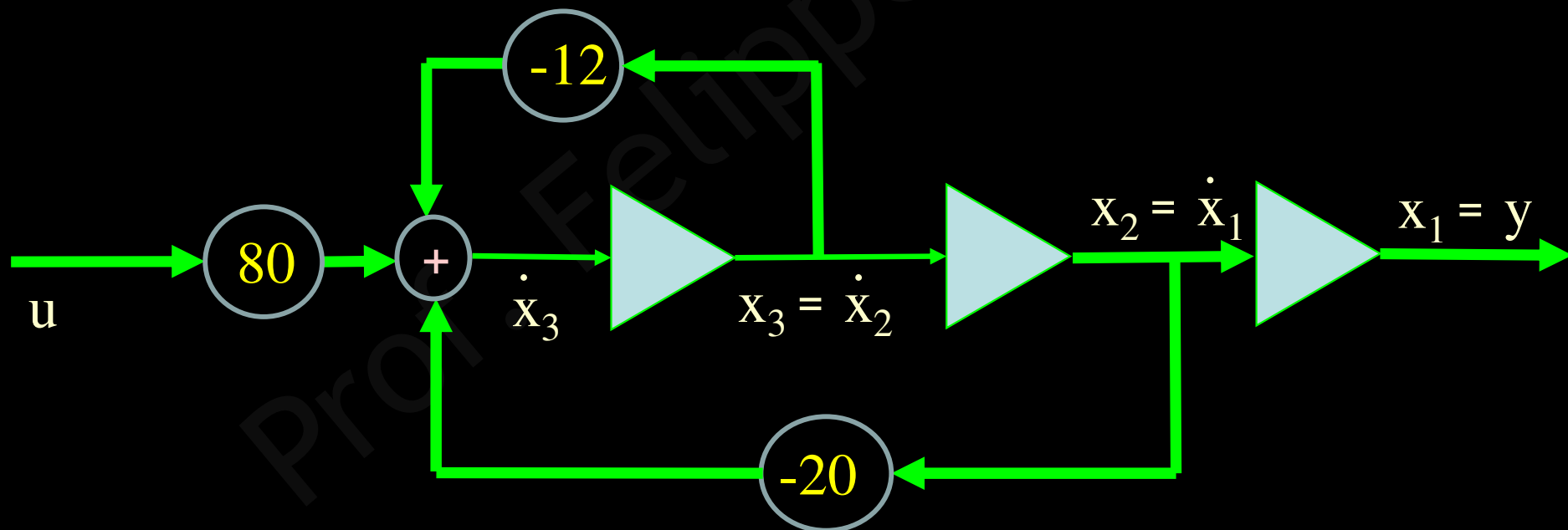
Note that **matrix A** is in the *companion form*



## Example 21:

Let us now do a *analog simulation* of this system A using the *state equation* obtained in the previous example

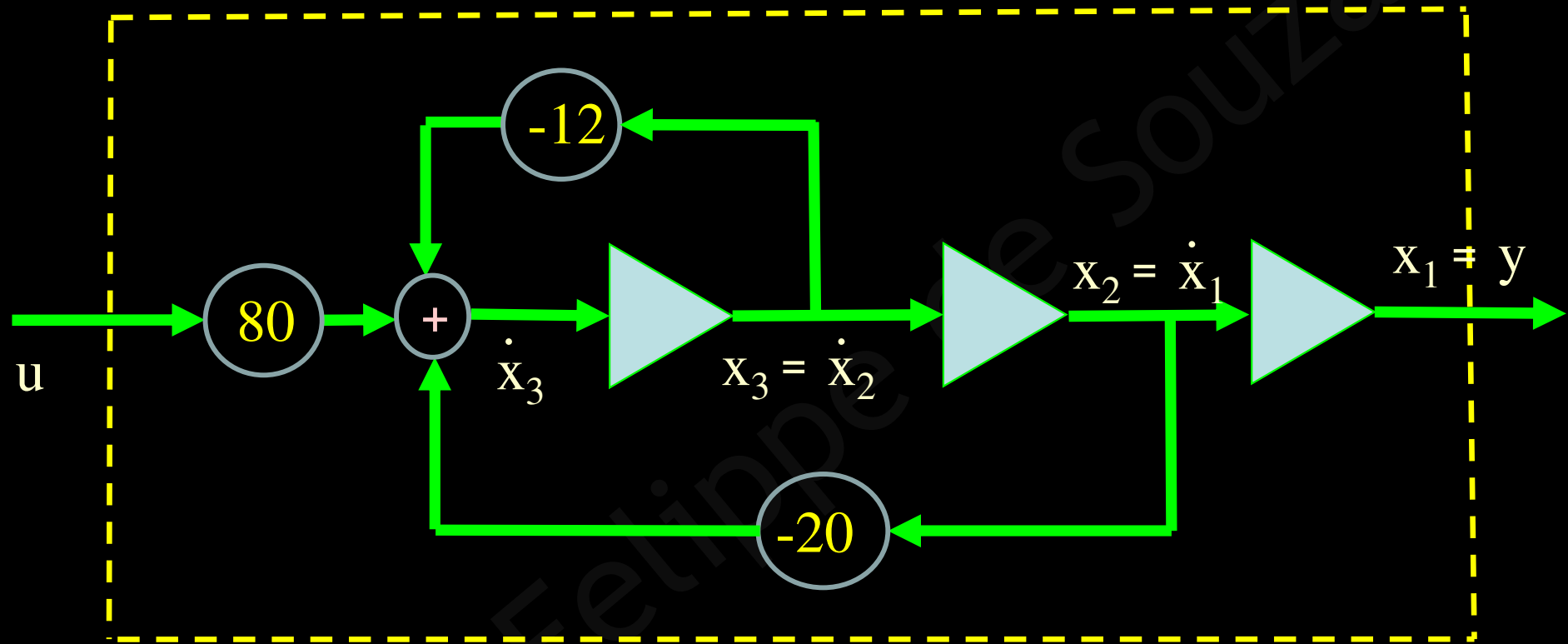
$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = x_3 \\ \dot{x}_3 = -20x_2 - 12x_3 + 80u \\ y = x_1 \end{cases}$$



# State Equations

## Example 20 (continued):

Now, by drawing a box covering the *analog simulation* done



we can observe that in this box only the *input*  $u$  comes in and only the *output*  $y$  comes out.

The variables  $x_1$  and  $x_2$  which stayed inside the box are part of an *internal representation* of the *system*, though its *state variable*

## Example 22:

Let us consider again the same system A of the previous example. However, here we are going to rewrite the *transfer function*  $G(s)$  given by (2) in the following form:

$$G(s) = \frac{5}{s} \cdot \frac{4}{(s+2)} \cdot \frac{4}{(s+10)} = \frac{Y(s)}{U(s)}$$


$\underbrace{\frac{5}{s} \cdot \frac{4}{(s+2)}}_{\frac{X_1(s)}{U(s)}}$


$\underbrace{\frac{4}{(s+10)}}_{\frac{X_2(s)}{U(s)}}$

$\underbrace{\frac{X_2(s)}{U(s)}}_{\frac{X_3(s)}{U(s)}}$


Defining the *state variable* in the following form

## Example 22 (continued):


$$\left\{ \begin{array}{l} X_1(s) = \frac{5 U(s)}{s} \\ X_2(s) = \frac{20 U(s)}{s(s+2)} \\ X_3(s) = G(s) \cdot U(s) = Y(s) = \frac{80 U(s)}{s(s+2)(s+10)} \end{array} \right.$$


$$\left\{ \begin{array}{l} sX_1(s) = 5U(s) \\ (s+2) X_2(s) = 4 \cdot \frac{5 U(s)}{s} = 4 \cdot X_1(s) \\ (s+10) X_3(s) = 4 \cdot \frac{20 U(s)}{s(s+2)} = 4 \cdot X_2(s) \end{array} \right.$$

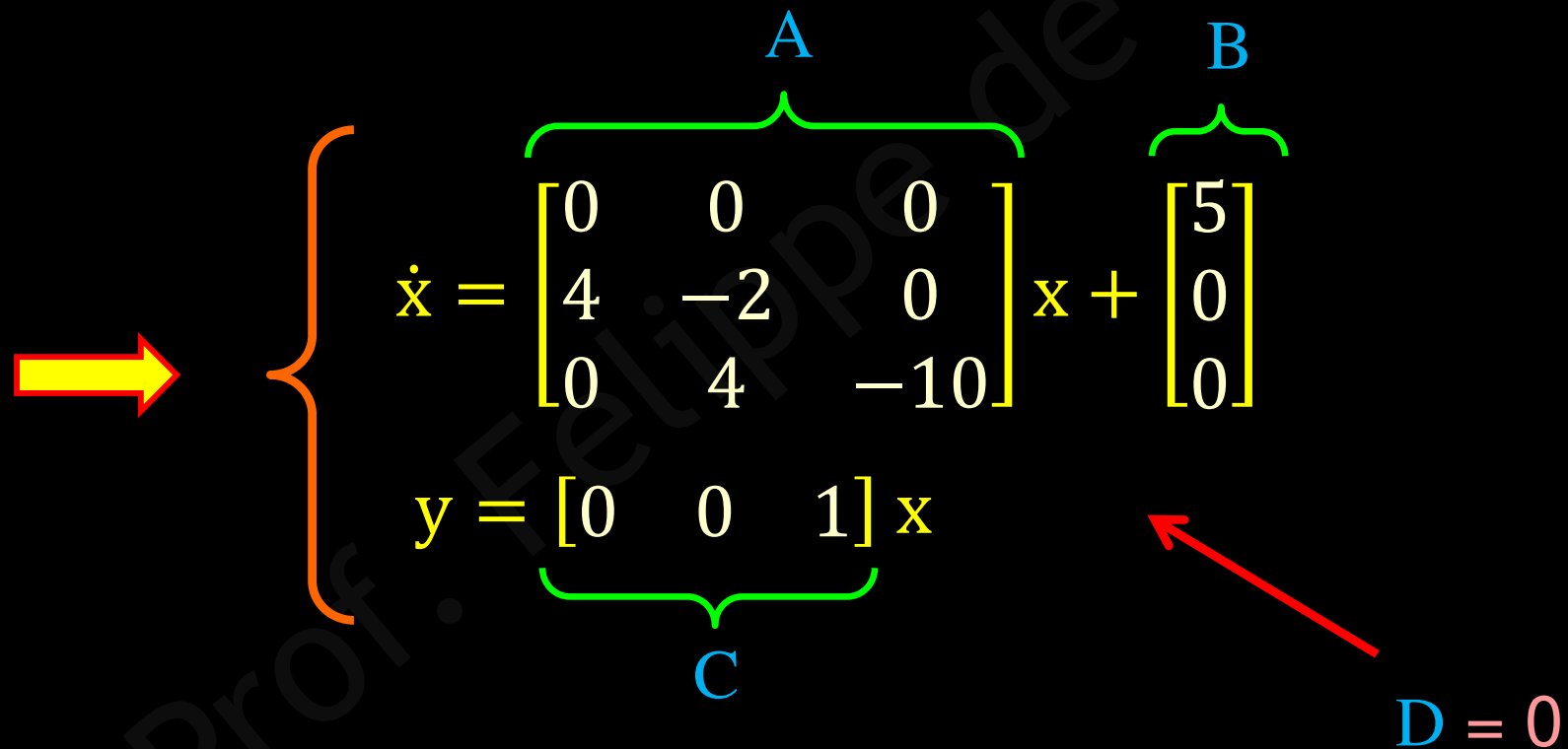
Example 22 (continued):


$$\left\{ \begin{array}{l} \dot{x}_1 = 5u \\ \dot{x}_2 = 4x_1 - 2x_2 \\ \dot{x}_3 = 4x_2 - 10x_3 \\ y = x_3 \end{array} \right.$$

Which give us a second *formulation* in *state equations* for this system A, different from the *formulation* of the previous *example*

## Example 22 (continued):

by writing in the **matrix form** we have



The diagram illustrates the matrix form of state equations. A large yellow arrow points to the left, indicating the transition to matrix form. The state equation is shown as  $\dot{x} = \begin{bmatrix} 0 & 0 & 0 \\ 4 & -2 & 0 \\ 0 & 4 & -10 \end{bmatrix} x + \begin{bmatrix} 5 \\ 0 \\ 0 \end{bmatrix}$ . The output equation is shown as  $y = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} x$ . The matrix  $\begin{bmatrix} 0 & 0 & 0 \\ 4 & -2 & 0 \\ 0 & 4 & -10 \end{bmatrix}$  is labeled **A** in blue. The vector  $\begin{bmatrix} 5 \\ 0 \\ 0 \end{bmatrix}$  is labeled **B** in blue. The vector  $\begin{bmatrix} 0 & 0 & 1 \end{bmatrix}$  is labeled **C** in blue. A red arrow points to the output equation, indicating that **D = 0**.

$$\dot{x} = \begin{bmatrix} 0 & 0 & 0 \\ 4 & -2 & 0 \\ 0 & 4 & -10 \end{bmatrix} x + \begin{bmatrix} 5 \\ 0 \\ 0 \end{bmatrix}$$
$$y = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} x$$

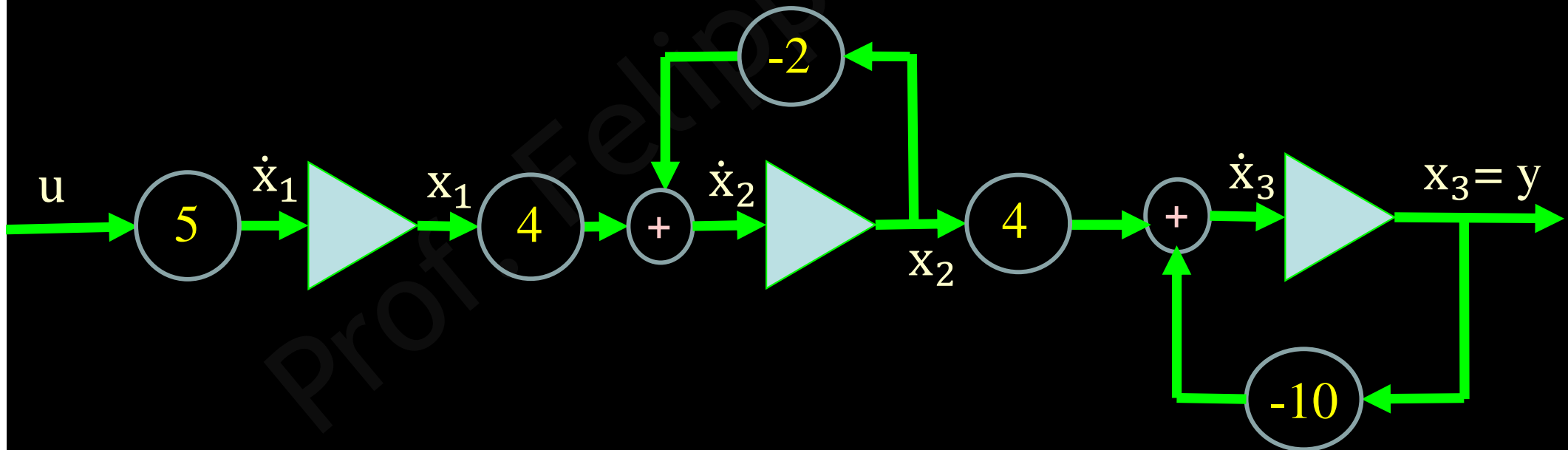
**D = 0**

# State Equations

## Example 23:

Let us now do a *analog simulation* of this system A using the *state equation* obtained in the previous example

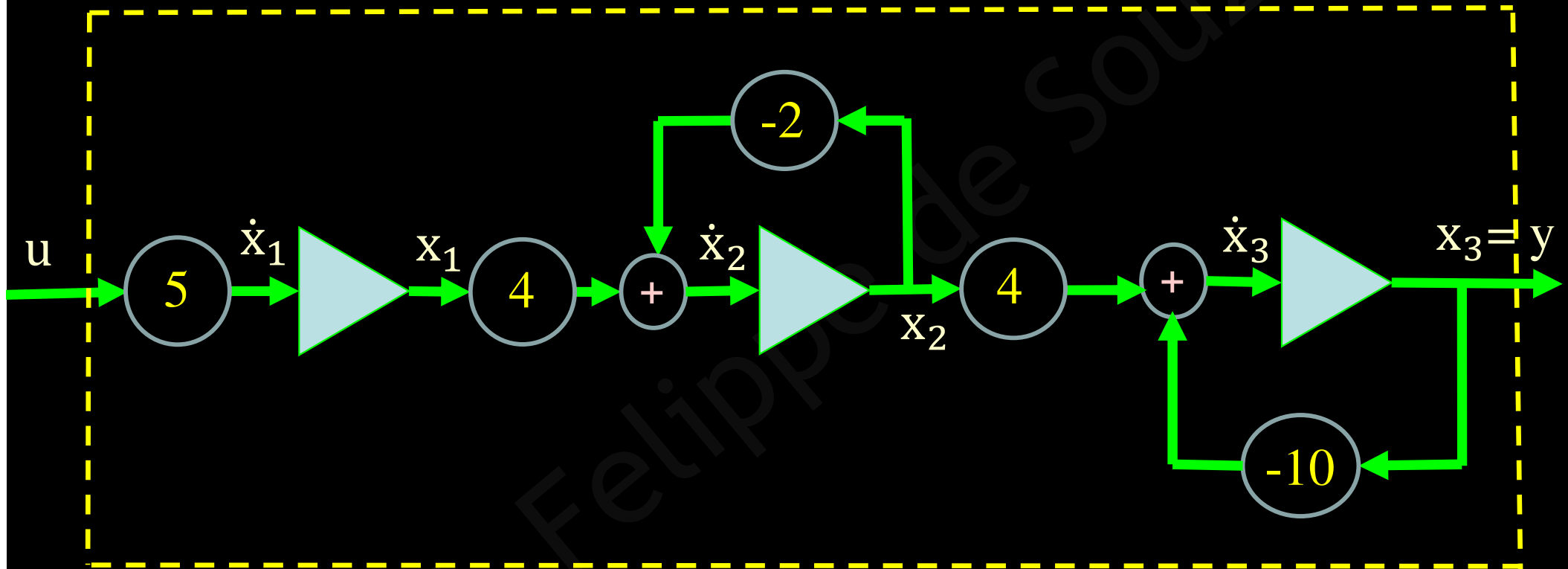
$$\begin{cases} \dot{x}_1 = 5u \\ \dot{x}_2 = 4x_1 - 2x_2 \\ \dot{x}_3 = 4x_2 - 10x_3 \\ y = x_3 \end{cases}$$



## State Equations

### Example 23 (continued):

Again, by drawing a box covering the *analog simulation* done



then we can observe that in this box only the *input*  $u$  comes in and only the *output*  $y$  comes out.

The variables  $x_1$  and  $x_2$  which stayed inside the box are part of an *internal representation* of the *system*, though its *state variable*



### Example 24:

Let us consider again the same system A of the 2 previous examples

However, here we are going to rewrite the *transfer function*  $G(s)$  given in (3) by expanding in *partial fractions* and defining the *state variables*  $X_1(s)$ ,  $X_2(s)$  and  $X_3(s)$  of the form shown below:

$$G(s) = \underbrace{\frac{4}{s}}_{\frac{X_1(s)}{U(s)}} + \underbrace{\frac{-5}{(s+2)}}_{\frac{X_2(s)}{U(s)}} + \underbrace{\frac{1}{(s+10)}}_{\frac{X_3(s)}{U(s)}} = \frac{Y(s)}{U(s)}$$


## Example 24 (continued):

we then have

$$\left\{ \begin{array}{l} X_1(s) = \frac{4U(s)}{s} \\ X_2(s) = \frac{-5U(s)}{(s+2)} \\ X_3(s) = \frac{U(s)}{(s+10)} \\ Y(s) = \left[ \frac{X_1(s)}{U(s)} + \frac{X_2(s)}{U(s)} + \frac{X_3(s)}{U(s)} \right] U(s) \end{array} \right. \rightarrow \left\{ \begin{array}{l} sX_1(s) = 4U(s) \\ sX_2(s) = -2X_2(s) - 5U(s) \\ sX_3(s) = -10X_3(s) + U(s) \\ Y(s) = X_1(s) + X_2(s) + X_3(s) \end{array} \right.$$


# State Equations

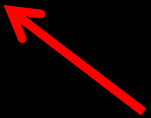
## Example 24 (continued):


$$\left\{ \begin{array}{l} \dot{x}_1 = 4u \\ \dot{x}_2 = -2x_2 - 5u \\ \dot{x}_3 = -10x_3 + u \\ y = x_1 + x_2 + x_3 \end{array} \right.$$

So, we have got a **third representation** in **state equations** for this same system A, different the previous ones.

Writing in **matrix form**


$$\left\{ \begin{array}{l} \dot{\mathbf{x}} = \overbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -10 \end{bmatrix}}^{\mathbf{A}} \mathbf{x} + \overbrace{\begin{bmatrix} 4 \\ -5 \\ 1 \end{bmatrix}}^{\mathbf{B}} u \\ y = \underbrace{\begin{bmatrix} 1 & 1 & 1 \end{bmatrix}}_{\mathbf{C}} \mathbf{x} \end{array} \right.$$

  $\mathbf{D} = 0$

# State Equations

Example 24 (continued):

$$\left\{ \begin{array}{l} \dot{\mathbf{x}} = \overbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -10 \end{bmatrix}}^{\mathbf{A}} \mathbf{x} + \overbrace{\begin{bmatrix} 4 \\ -5 \\ 1 \end{bmatrix}}^{\mathbf{B}} u \\ y = \underbrace{[1 \quad 1 \quad 1]}_{\mathbf{C}} \mathbf{x} \end{array} \right. \quad \mathbf{D} = 0$$

Note that *matrix*  $\mathbf{A}$  is in the diagonal form in this *representation* and the *poles* of the system ( $s = 0$ ,  $s = -2$  and  $s = -10$ ) are the elements of the *main diagonal*

It is obvious that this happens: since *matrix*  $\mathbf{A}$  is diagonal, then the *elements* of its *main diagonal* are the own *eigenvalues* of the *system*.

# State Equations

## Example 25:

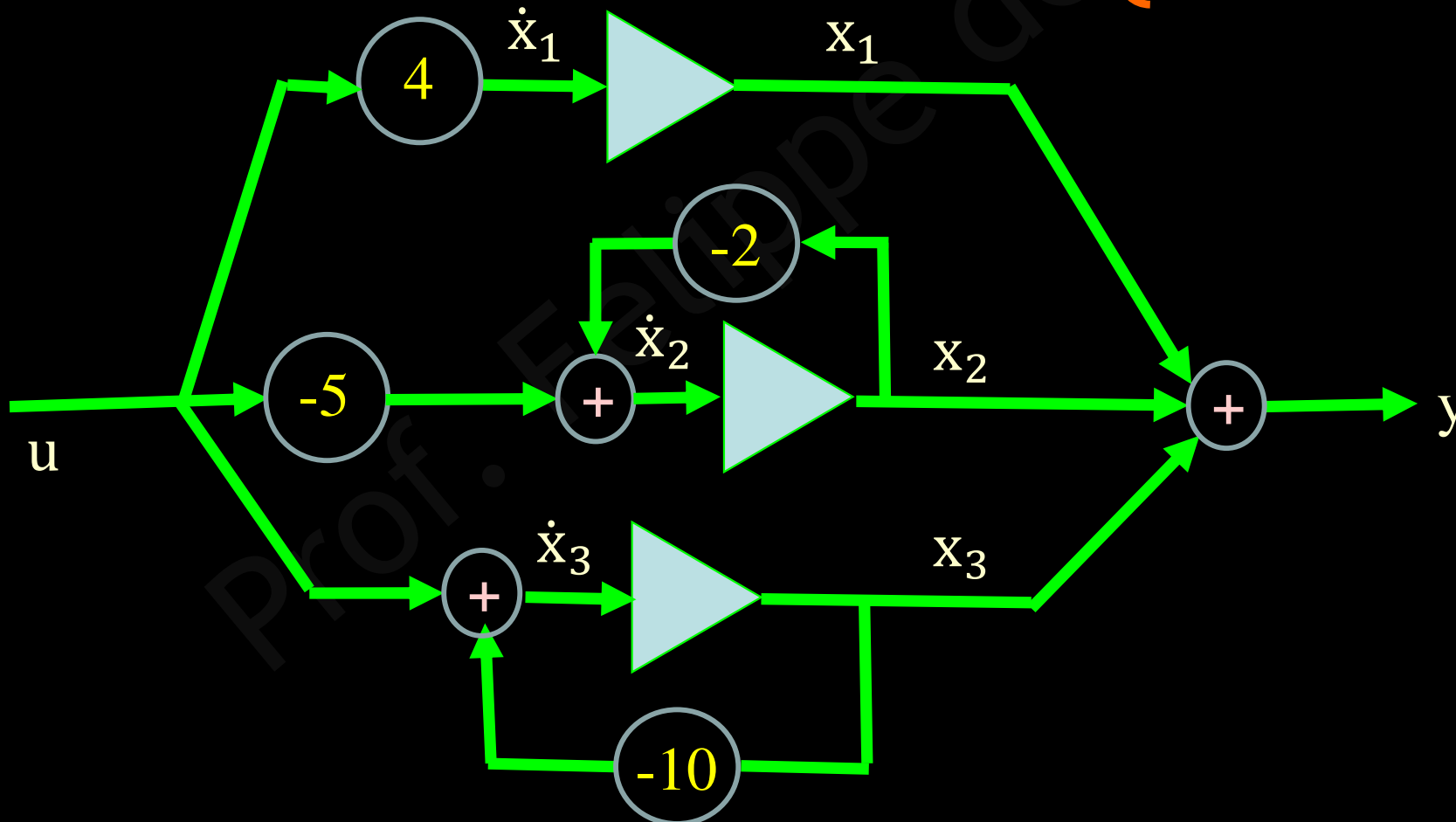
Let us now do a *analog simulation* of this system A using the *state equation* obtained in the previous example

$$\dot{x}_1 = 4u$$

$$\dot{x}_2 = -2x_2 - 5u$$

$$\dot{x}_3 = -10x_3 + u$$

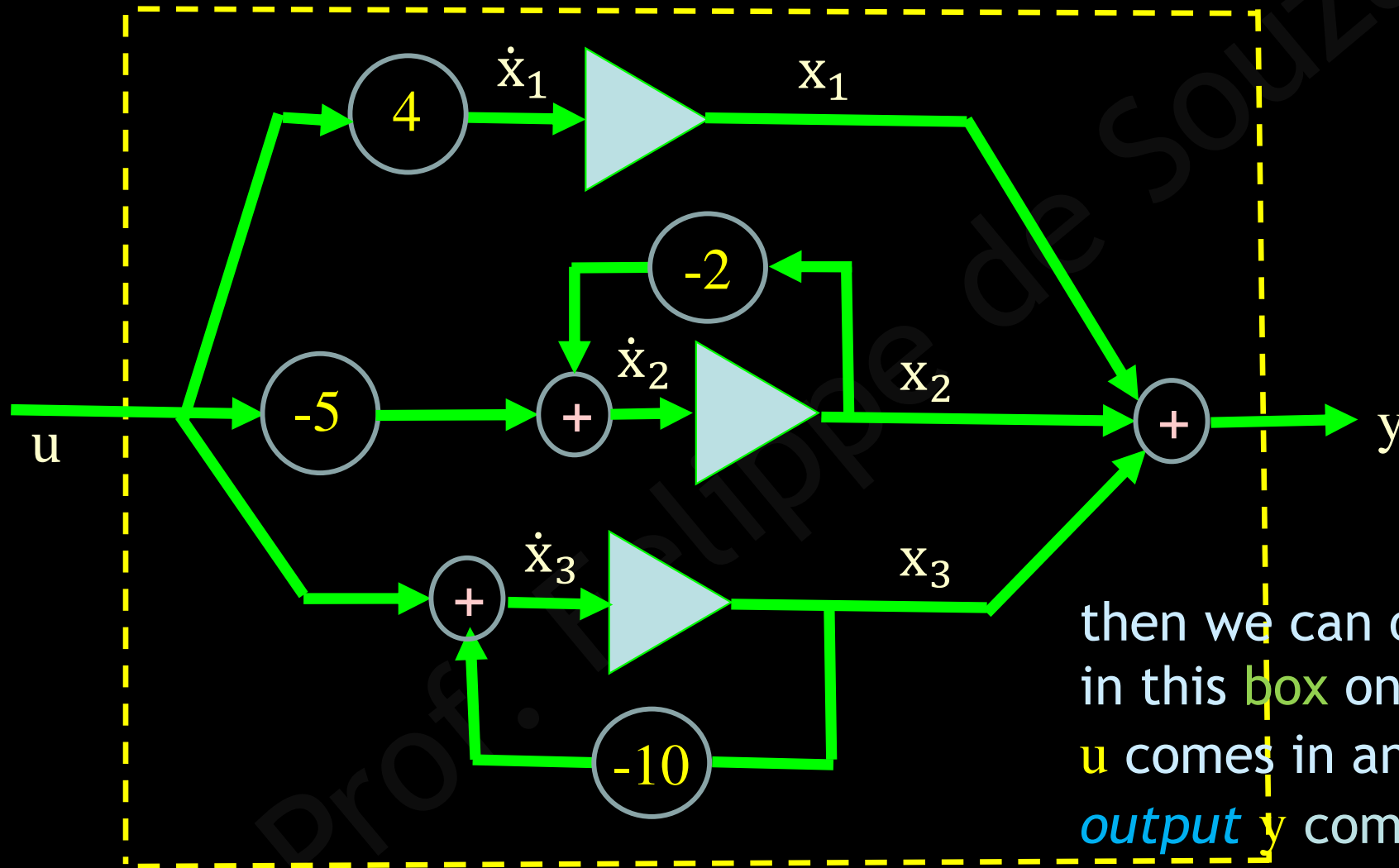
$$y = x_1 + x_2 + x_3$$



# State Equations

## Example 25 (continued):

Once again, by drawing a box covering the *analog simulation* done



then we can observe that in this box only the *input*  $u$  comes in and only the *output*  $y$  comes out.

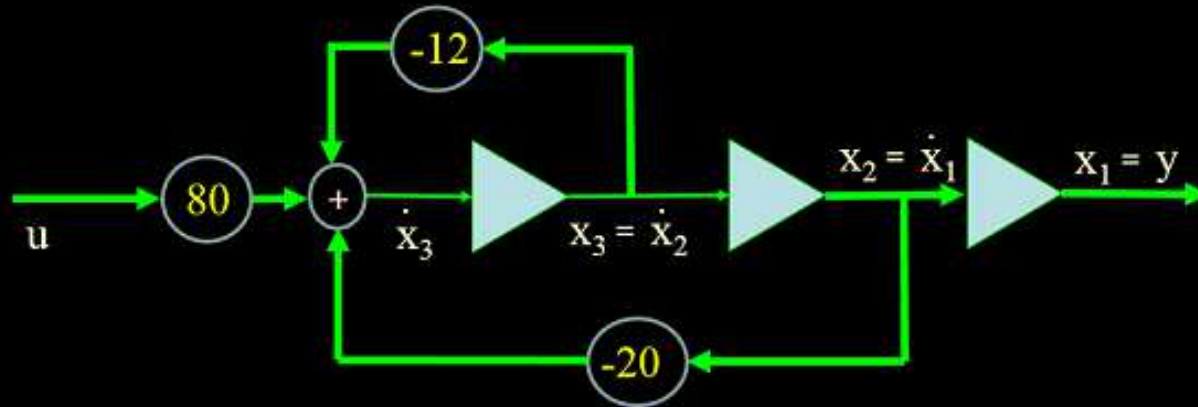
The variables  $x_1$  and  $x_2$  which stayed inside the box are part of an *internal representation* of the *system*, though its *state variable*.

# State Equations

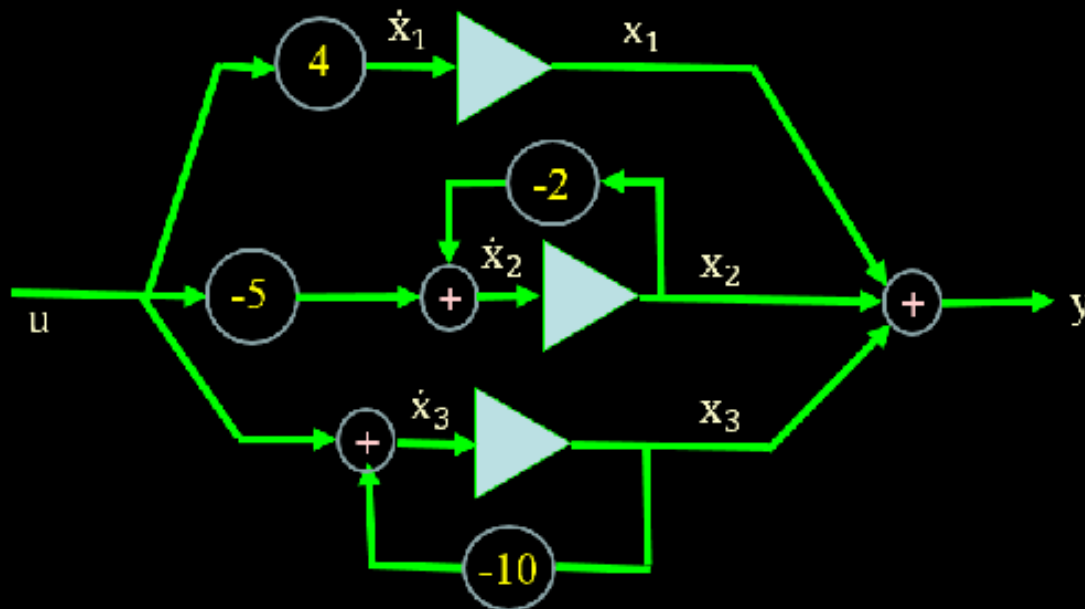
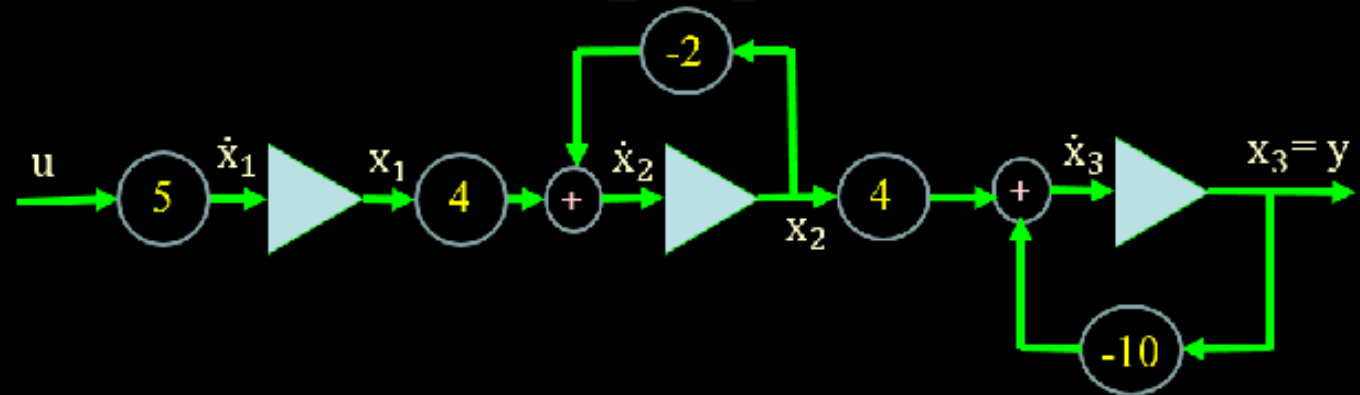
In the previous **examples** we have obtained **3** different *representations* in *state equations* for the same *system*

$$\left\{ \begin{array}{l} \dot{x}_1 = x_2 \\ \dot{x}_2 = x_3 \\ \dot{x}_3 = -20x_2 - 12x_3 + 80u \\ y = x_1 \end{array} \right. \quad \left\{ \begin{array}{l} \dot{x}_1 = 5u \\ \dot{x}_2 = 4x_1 - 2x_2 \\ \dot{x}_3 = 4x_2 - 10x_3 \\ y = x_3 \end{array} \right.$$
$$\left\{ \begin{array}{l} \dot{x}_1 = 4u \\ \dot{x}_2 = -2x_2 - 5u \\ \dot{x}_3 = -10x_3 + u \\ y = x_1 + x_2 + x_3 \end{array} \right.$$

# State Equations



as well as we have obtained **3 different analog simulations** for the same *system*





## State Equations

As we have already seen in the sections

*“Equivalent representations”,*

the representation of a system in *state equations*

It is not unique!

If the *state variable* is  $\mathbf{x}(t)$ , then for every invertible matrix  $\mathbf{P}$ , we can get a new *state variable*

$$\bar{\mathbf{x}}(t) = \mathbf{P} \mathbf{x}(t)$$

and thus, a new  
*representation* of the  
*system* in *state equations*

$$\left\{ \begin{array}{l} \dot{\bar{\mathbf{x}}} = \bar{\mathbf{A}} \bar{\mathbf{x}} + \bar{\mathbf{B}} \mathbf{u} \\ \mathbf{y} = \bar{\mathbf{C}} \bar{\mathbf{x}} + \bar{\mathbf{D}} \mathbf{u} \end{array} \right.$$



Departamento de  
Engenharia Eletromecânica

Thank you!  
Obrigado!

Felippe de Souza

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